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**Estimation and Identification in Dynamic Systems (048825)**

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Lecture Notes

\* Revised version, incorporating updates by Prof. Ron Meir



# 1 Introduction: Basic Problems of Interest

Our main interest in this course will be in the following problem:

- State estimation in dynamic systems, for which the state cannot be fully observed.

Some related problems that we shall also consider:

- Parameter estimation in dynamic systems (system identification).
- Joint state and parameter estimation.

Our emphasis will be on algorithms which are optimal in a statistical (stochastic) sense.

The basic system models that we will deal with are:

- Continuous-state systems: We will develop the celebrated *Kalman Filter* for state estimation in linear state-space models, and its extensions.
- Discrete-state models: the so-called Hidden Markov Models.

We will also consider various extensions of these basic models and problems.

We next give a brief outline of the basic problems and illustrative applications.

## 1.1 State Estimation in Linear Systems

Consider a discrete-time linear state-space model of the form:

$$\begin{aligned}x_{k+1} &= Fx_k + Gu_k + v_k, \\z_k &= Hx_k + w_k.\end{aligned}$$

Here:  $x \in \mathbb{R}^n$  is the state vector, which is unknown to us

$z \in \mathbb{R}^m$  is the measurement vector

$u$  is a known input signal

$v$  and  $w$  are unobserved noise sequences

$F, G, H$  are the system matrices

*The basic state-estimation problem:* The systems matrices are given, and so are some properties of the noise sequences. Our goal is to find an estimate  $\hat{x}_{k+1}$  for the state vector  $x_{k+1}$ , given the measurements  $\{z_k, z_{k-1} \dots\}$ .

The proposed solution is of the following (state-observer) form:

$$\hat{x}_{k+1} = F\hat{x}_k + Gu_k + K_k(z_k - H\hat{x}_k).$$

This is a **recursive filter**, which can be operated in an on-line mode (i.e., the estimate is updated each time a new measurement is obtained).

$K_k$  is a gain matrix, to be “properly” chosen.

The Kalman Filter is obtained by an optimal choice of these gains, under appropriate statistical model assumptions and error criteria.

**Extensions:** Most models of interest in practice are non-linear. Various (sub-optimal) extensions of the Kalman filter to non-linear problems now exist and are widely used, we will cover these as well.

**Applications:** The Kalman filter is routinely used in navigation systems (own position, velocity and acceleration estimation), tracking systems (same, for other objects), control systems (state estimation for state feedback control), along with numerous other applications in signal processing and related fields.

*Examples:* We next sketch a few simplified examples for problems that can be cast in this form.

### Example 1: Position Estimation

Consider an object moving in 1-dimensional space, with position  $p(t)$ .

We are given *noisy* (inaccurate) measurements of this position at some discrete times:

$$z(t) = p(t) + n_z(t), \quad t = t_0, t_1, \dots$$

Required: to estimate the position  $\hat{p}(t)$ .

To formulate this problem in state space form, several options are available:

1. Random acceleration model (2nd order model):

$$\begin{aligned} \frac{d}{dt}p(t) &= v(t) \\ \frac{d}{dt}v(t) &= a(t) \equiv n_v(t) \end{aligned}$$

where  $n_v(t)$  is “white”, 0-mean noise signal with known statistics. This noise reflects the expected object “maneuverability”.

We have arrived at the following state model:

$$\frac{d}{dt} \begin{bmatrix} p(t) \\ v(t) \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} p(t) \\ v(t) \end{bmatrix} + \begin{bmatrix} 0 \\ n_v(t) \end{bmatrix}$$

where  $x(t) = (p(t), v(t))'$  is the state vector.

This state equation may be discretized to obtain a discrete-time state model over the measurement times, of the form:

$$x(t_{k+1}) = A(k)x(t_k) + n(k).$$

The measurement equation is

$$z(t) = [1, 0] \begin{bmatrix} p(t) \\ v(t) \end{bmatrix} \equiv [1, 0] x(t) \quad t = t_k$$

2. Random acceleration-change model (3rd order model):

When velocity cannot change abruptly, the following model is more suitable:

$$\begin{aligned}\dot{p}(t) &= v(t) \\ \dot{v}(t) &= a(t) \\ \dot{a} &= n_a(t)\end{aligned}$$

Furthermore, when acceleration cannot change abruptly we can add a simple low path filter:

$$\dot{a} = -\beta a + n_a(t)$$

with  $\beta$  a properly chosen constant. With state  $x(t) = [p(t), v(t), a(t)]'$ , we obtain the following state model:

$$\dot{x}(t) = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & -\beta \end{bmatrix} x(t) + \begin{bmatrix} 0 \\ 0 \\ n_a(t) \end{bmatrix}$$

and measurement equation:

$$z = [1, 0, 0]x + n_z.$$

3. Additional measurements:

We may, for example, have also direct velocity measurements of the moving object. Then the measurement equations are:

$$z_1(t_k) = p(t_k) + n_1(t_k) \quad z_2(t_k) = v(t_k) + n_2(t_k)$$

In matrix form (for  $x = [p, v, a]'$ ):

$$z(t_k) = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix} x(t) + n_z(t)$$

with  $n_z = (n_1, n_2)'$ .

Position estimation (with Kalman filtering) has many variants and applications, including:

- *Navigation*: Using odometry, inertial sensors, GPS, vision,...
- *Tracking*: Using radar, vision, cellular phones....

### Example 2: Signal Detection

Consider a discrete-time signal  $s(k)$  which is transmitted through a noisy channel with ISI; the received signal is

$$z(k) = \sum_{i=0}^{N-1} h_i s(k-i) + n_z(k)$$

with  $n_z$  a white noise sequence, say  $n_z(k) \sim N(0, \sigma_z)$ .

It is required to recover the transmitted signal  $s(k)$  from the measurements  $z(k')$ ,  $k' \leq k$ . This is a classical signal filtering problem.

To use statistical methods, we use a statistical model for the transmitted signal: e.g.,  $s$  is a white noise sequence with  $s(k) \sim N(0, \sigma_s)$ . For  $N = 2$ , the state variables and equations are:

$$\begin{aligned} x_1(k) = s(k) & \Rightarrow x_1(k+1) = n_s(k+1) \\ x_2(k) = s(k-1) & \Rightarrow x_2(k+1) = x_1(k) \\ x_3(k) = s(k-2) & \Rightarrow x_3(k+1) = x_2(k) \end{aligned}$$

and in matrix form:

$$x(k+1) = \begin{bmatrix} 0 & 0 & 0 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix} x(k) + \begin{bmatrix} n_s(k+1) \\ 0 \\ 0 \end{bmatrix}$$

$$z(k) = [h_0, h_1, h_2]x(k) + n_z(k).$$

### Example 3: Higher harmonics detection

We are given a sinusoidal signal, with basic frequency  $f_1 = 50$ . We need to detect the higher-order (say, 3rd order) harmonic content of the signal. (A 2nd order harmonic is usually absent in power systems.)

With  $\omega_1 = 2\pi f_1$ , we write this signal as

$$s(t) = A \sin(\omega_1 t + \phi_1) + B \sin(3\omega_1 t + \phi_3) \text{ (+other terms).}$$

It is assumed that the amplitude and phases may (slowly) vary with time. The signal is measured every  $T = 0.1$  sec. We wish to track  $B(t)$  and  $\phi_1(t)$  over time.

The problem may obviously be approached in the frequency domain, using standard filtering methods. We give here the alternative Kalman-filter formulation.

We start by writing the harmonic signal model as:

$$s(t) = A_1(t) \cos(\omega_1 t) + A_2(t) \sin(\omega_1 t) + B_1(t) \cos(3\omega_1 t) + B_2(t) \sin(3\omega_1 t)$$

The state is taken as the four amplitudes:  $x = [A_1, A_2, B_1, B_2]'$ . Focusing on the measurements instances  $t_k = kT$ , we have the following model:

$$\begin{aligned} A_1(t_{k+1}) &= A_1(t_k) + n_{A1}(k) \\ &\vdots \\ B_2(t_{k+1}) &= B_2(t_k) + n_{B2}(k) \end{aligned}$$

The noise variances are taken as small quantities related to the allowed rate of change of the amplitudes. The state equation is then

$$x(t_{k+1}) = Ix(t_k) + n(k)$$

where  $I$  is the unit matrix.

The measurement model is:

$$z(t_k) = s(t_k) + n_z(k)$$

with  $n_z$  the measurement error. It may be taken as a white Gaussian sequence, with  $n_z(k) \sim N(0, \sigma_z)$ . This gives

$$\begin{aligned} z(t_k) &= [\cos(\omega_1 t_k), \sin(\omega_1 t_k), \cos(3\omega_1 t_k), \sin(3\omega_1 t_k)]x(t_k) + n_z(k) \\ &\doteq H(k)x(t_k) + n_z(k). \end{aligned}$$

## 1.2 Hidden Markov Models (HMMs)

HMMs are state models with discrete state, which cannot be directly observed, and with discrete or continuous measurements.

Let  $x_k \in \{1, 2, \dots, N\}$  be a Markov chain specified by the transition law

$$p(x_{k+1} = j | x_k = i) = p_{ij}$$

and initial distribution  $p(x_0)$ . Let the  $z_k$  be the measurement, say discrete, related to  $x_k$  through,

$$p(z_k = z | x_k = i) = q(z|i).$$

The basic problems here are:

1. Given the model parameters, and measurements  $(z_n, z_{n-1}, \dots, z_1)$ , estimate the state sequence  $\{x_n, x_{n-1}, \dots, x_1\}$ .

We shall develop the Maximum Likelihood (ML) estimator for that problem, which is efficiently implemented using the so-called *Viterbi algorithm*.

2. Given the measurements  $(z_n, z_{n-1}, \dots, z_1)$ , estimate the model parameters, namely: find the model that best describes the data.

The standard solution for this problem is joint state and parameter estimation, using the *EM* (Expectation Maximization) algorithm.

**HMM Applications:** HMMs are a basic tool for pattern recognition (or machine learning) in temporal or sequential data. Major application areas include speech and language processing, computational biology (bioinformatics), gesture recognitions, along with many others.

Consider the application to speech processing, in particular speech recognition. The HMM may be used at different levels of speech modeling, such as:

*Word level:* The state  $x$  is a complete word; the measurement is the recorded sound of the word; and the dynamics  $p(j|i)$  represents the likelihood of word  $j$  appearing after word  $i$ .

*Phonetic level:* Using a phonetic alphabet to model the inner structure of each word.

In that context, “estimating the state sequence” means identifying the spoken word sequence; and “estimating the model parameters” relates to the training phase when the model is tuned to a specific speaker.